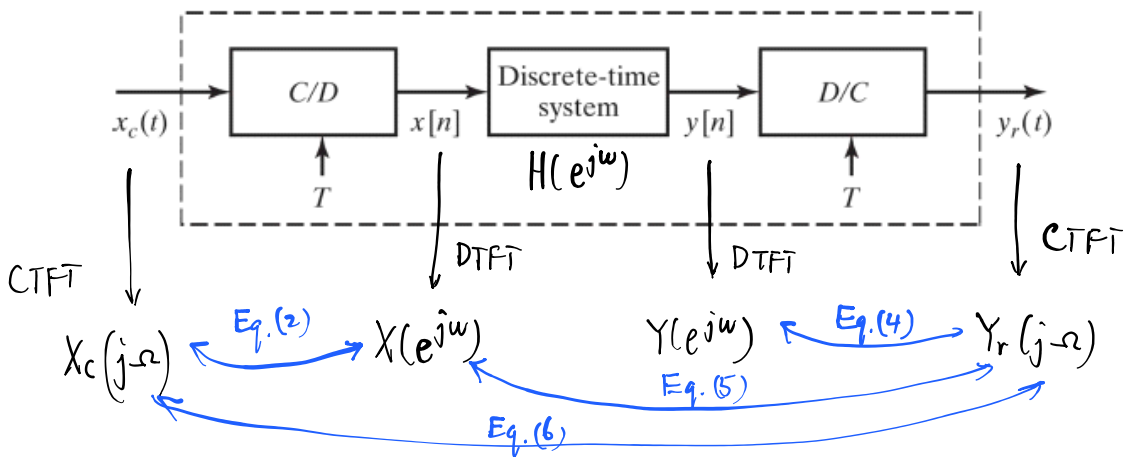


Lecture 11

Discrete-Time Processing of Continuous-Time Signals



- C/D Converter

$$x[n] = x_c(t) \Big|_{t=nT} = x_c(nT) \quad \dots \dots (1)$$

$$X(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\frac{\omega}{T} - \frac{2\pi k}{T})) \quad \dots \dots (2)$$

- D/C Converter

$$y_r(t) = \sum_{n=-\infty}^{\infty} y[n] \cdot \frac{\sin[\frac{\pi(t-nT)}{T}]}{\frac{\pi(t-nT)}{T}} \quad \dots \dots (3)$$

$$Y_r(j-\Omega) = H_r(j-\Omega) \cdot Y(e^{j\Omega T}) \quad \dots \dots (4)$$

- D-T system has a system function $H(e^{j\omega})$

$$Y(e^{j\omega}) = H(e^{j\omega}) X(e^{j\omega})$$

From (4): $Y_r(j-\Omega) = H_r(j-\Omega) \cdot H(e^{j\Omega T}) \cdot X(e^{j\Omega T}) \quad \dots \dots (5)$

From (2) :

$$X(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\frac{\omega}{T} - \frac{2\pi k}{T})) \dots\dots (2)$$

$$\begin{aligned} X(e^{j\Omega T}) &= \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\frac{\Omega T}{T} - \frac{2\pi k}{T})) \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - \frac{2\pi k}{T})) \end{aligned}$$

From (5) :

$$Y_r(j\Omega) = H_r(j\Omega) \cdot H(e^{j\Omega T}) \cdot X(e^{j\Omega T}) \dots\dots (5)$$

$k=0$
 $X_c(j\Omega)$: original spectrum

$$Y_r(j\Omega) = H_r(j\Omega) \cdot H(e^{j\Omega T}) \cdot \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - \frac{2\pi k}{T})) \dots\dots (6)$$

Since

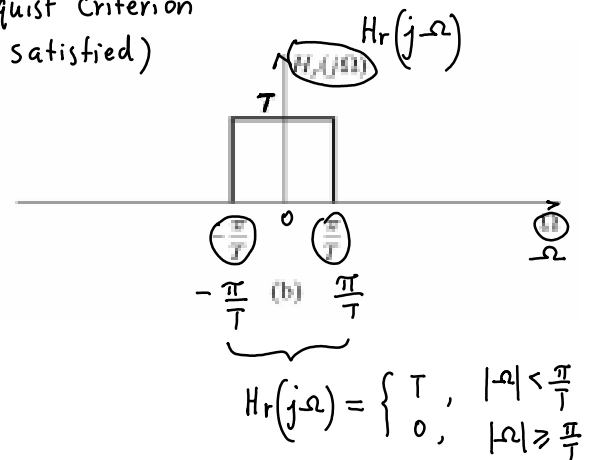
$$X(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\frac{\omega}{T} - \frac{2\pi k}{T})) \dots\dots (2)$$

Assumption : The input $x_c(t)$ is bandlimited

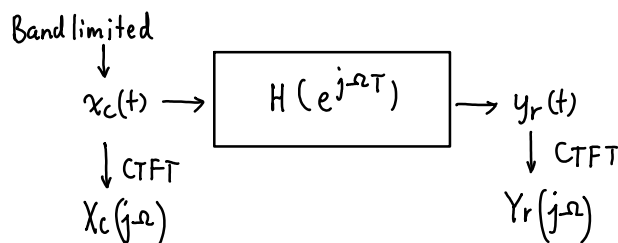
$$X_c(j\Omega) = 0 \quad \text{if} \quad |\Omega| \geq \frac{\pi}{T} \quad (\text{Nyquist criterion is satisfied})$$

Then from (6)

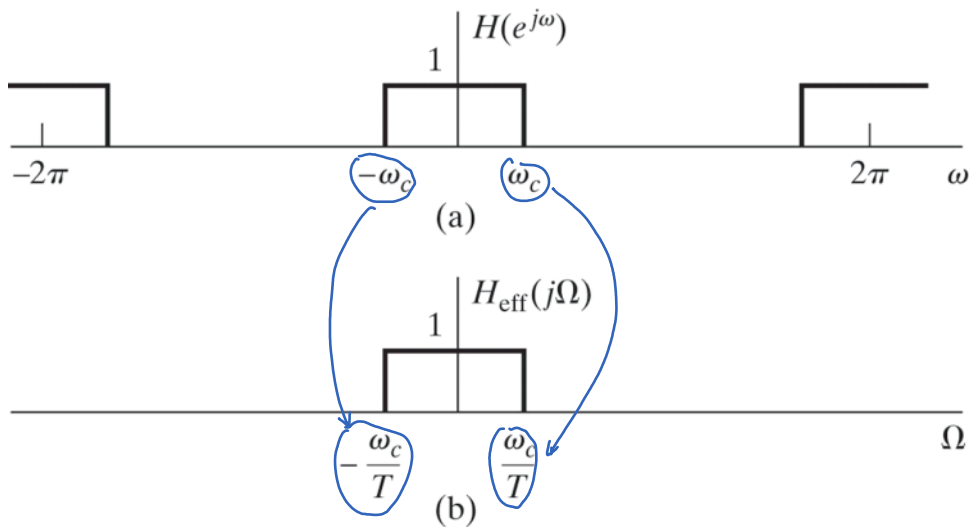
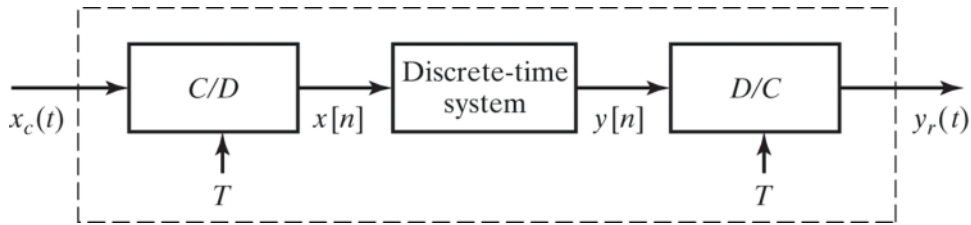
$$Y_r(j\Omega) = \begin{cases} 0, & |\Omega| \geq \frac{\pi}{T} \\ H(e^{j\Omega T}) X_c(j\Omega), & |\Omega| < \frac{\pi}{T} \end{cases}$$



The overall continuous-time system is equivalent to an LTI system:



$H_{\text{eff}}(j\Omega)$: CTFT of the entire system



$$Y_r(j\Omega) = \begin{cases} 0, & |\Omega| \geq \frac{\pi}{T} \\ H(e^{j\Omega T})X_c(j\Omega), & |\Omega| < \frac{\pi}{T} \end{cases}$$

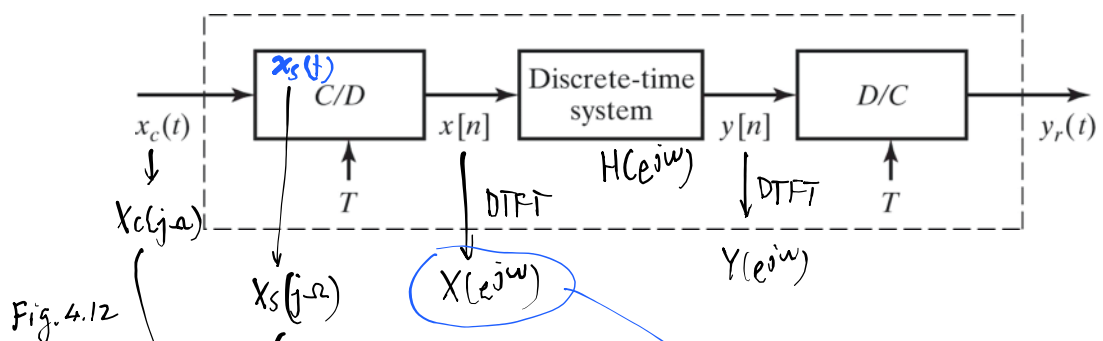
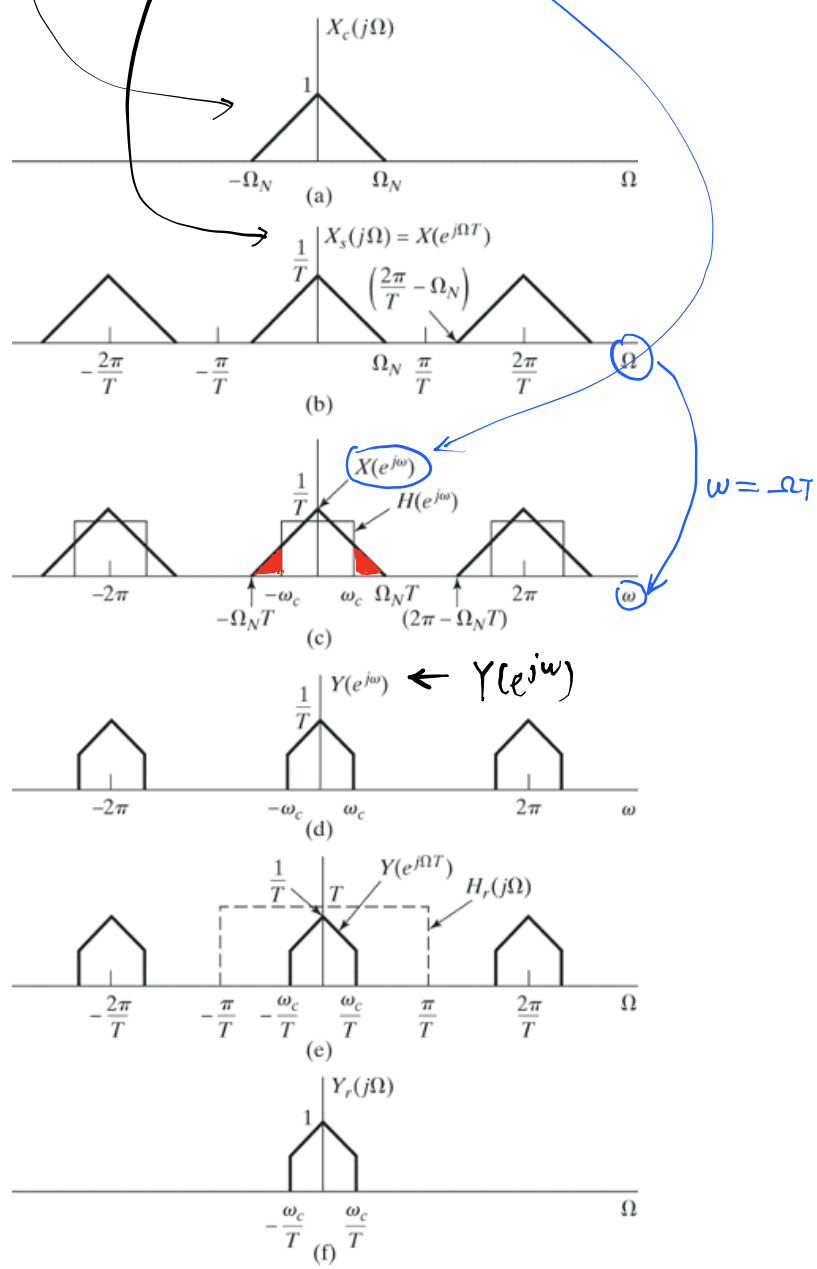
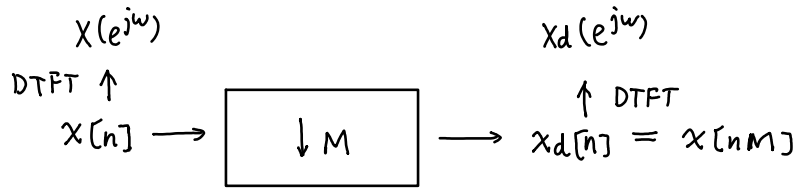


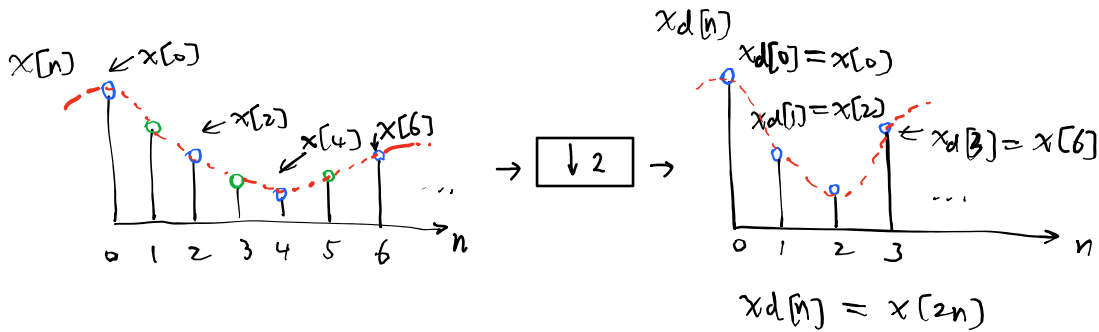
Fig. 4.12



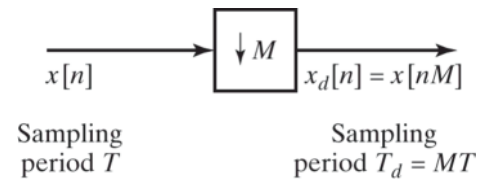
- Change the Sampling rate using Discrete-Time Processing



For example, $M = 2$



In frequency domain, suppose that there exists $X_c(f)$



$$x[n] = x_c(nT)$$

$$\downarrow \text{DTFT}$$

$$X(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c\left(j\left(\frac{\omega}{T} - \frac{2\pi k}{T}\right)\right) \quad \dots \dots E_{\frac{1}{T}}(z)$$

$$x_d[n] = x_c(nMT)$$

$$\downarrow \text{DTFT}$$

$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_{k=-\infty}^{\infty} X_c\left(j\left(\frac{\omega}{MT} - \frac{2\pi k}{MT}\right)\right)$$

change the variable $k \rightarrow r$, rewrite $X_d(e^{j\omega})$

$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_{r=-\infty}^{\infty} X_c\left(j\left(\frac{\omega}{MT} - \frac{2\pi r}{MT}\right)\right)$$

Let $r = i + kM$, partitioning of r such that $\begin{cases} k: \text{quotient} \\ i: \text{remainder} \end{cases}$ of $\frac{r}{M}$.

For example, if $M=2$, then $r = i + 2 \cdot k \rightarrow \frac{r}{2} \rightarrow \begin{cases} k: \text{quotient} \\ i: \text{remainder} \end{cases}$

	r	i	k
even \rightarrow	0	0	0
odd \rightarrow	1	①	0
	2	0	1
	3	①	1
	4	0	2
	5	①	2
	\vdots	\vdots	\vdots

Go back to $X_d(e^{j\omega})$

$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_{r=-\infty}^{\infty} X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi r}{MT} \right) \right)$$

$$r = i + kM$$

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} \left[\frac{1}{T} \sum_{k=-\infty}^{\infty} X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi kM}{MT} - \frac{2\pi i}{MT} \right) \right) \right]$$

How about

$$X(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c \left(j \left(\frac{\omega}{T} - \frac{2\pi k}{T} \right) \right) \dots \text{Eq. (2)}$$

$$X \left(e^{j \underbrace{\left(\frac{\omega - 2\pi i}{M} \right)}} \right) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c \left(j \left(\frac{\omega - 2\pi i}{MT} - \frac{2\pi k}{T} \right) \right)$$

Thus $\frac{\omega}{M} - \frac{2\pi i}{M}$

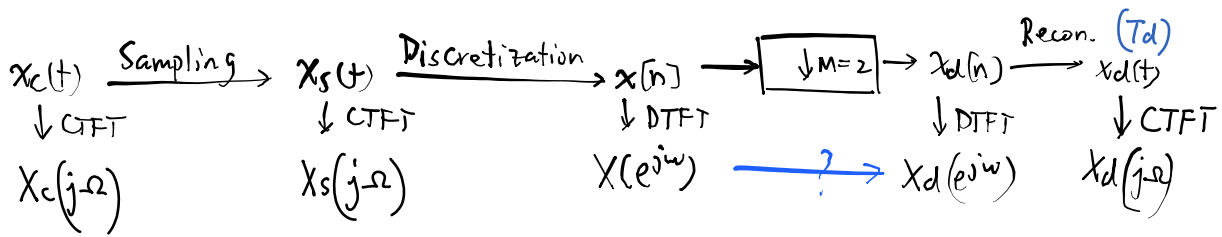
$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X \left(e^{j \left(\frac{\omega}{M} - \frac{2\pi i}{M} \right)} \right)$$

$\frac{\omega}{M}$: scaling

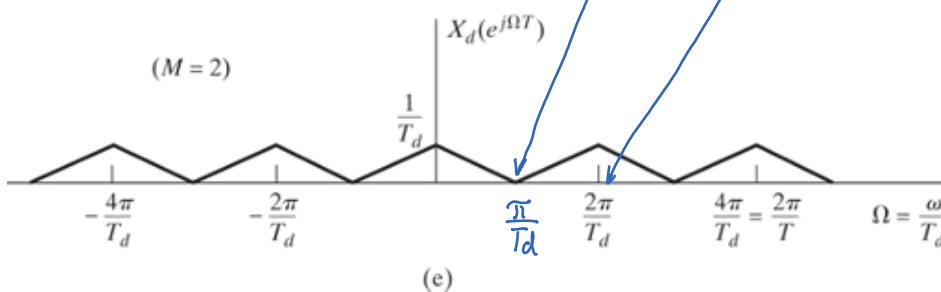
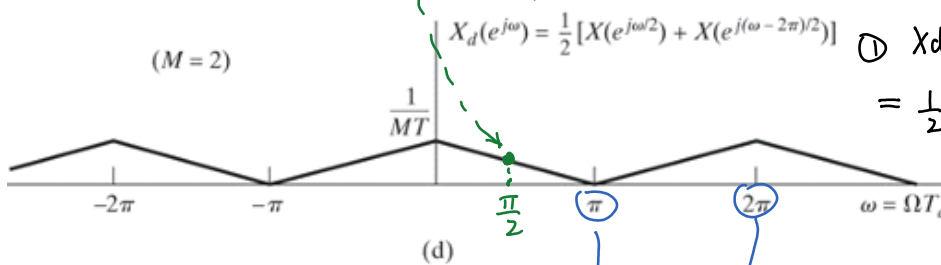
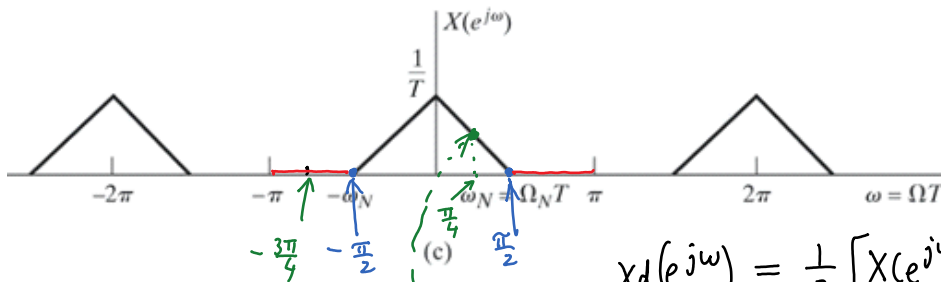
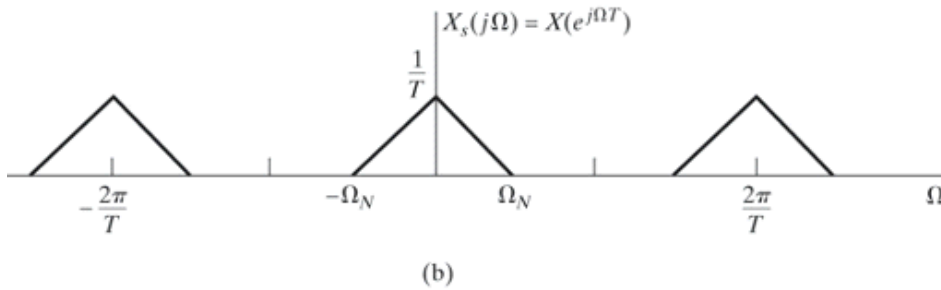
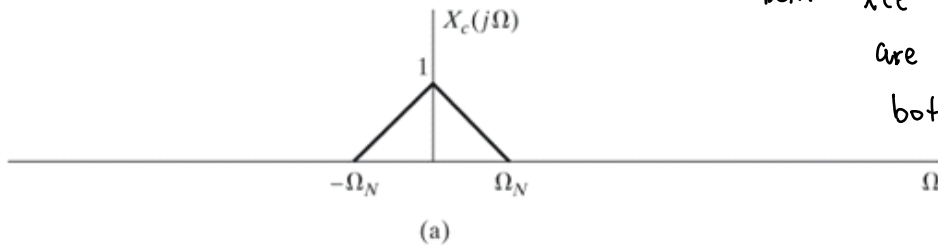
$-\frac{2\pi i}{M}$: shifting

$X_d(e^{j\omega})$: M copies of the DTFT of $x[n]$, $X(e^{j\omega})$ with the frequency scaled by M and shifted by integer multiples of 2π .

$\underbrace{\hspace{10em}}_{\text{expansion}}$



Both $X(e^{j\omega})$ and $X_d(e^{j\omega})$ are DTFT's, so they are both periodic with fundamental period $= 2\pi$.



$$x_d(e^{j\omega}) = \frac{1}{2} [X(e^{j\omega/2}) + X(e^{j(\frac{\omega}{2} - \pi)})]$$

$$X_d(e^{j\omega}) = \frac{1}{2} [X(e^{j\omega/2}) + X(e^{j(\frac{\omega}{2} - \pi)})]$$

(M=2)

$$\textcircled{1} X_d(e^{j\pi}) \text{ if } \omega = \pi$$

$$= \frac{1}{2} [X(e^{j\frac{\pi}{2}}) + X(e^{j(\frac{\pi}{2} - \pi)})]$$

$$= \frac{1}{2} [0 + 0] = 0$$

$$\textcircled{2} X_d(e^{j\frac{\pi}{2}}) \text{ if } \omega = \frac{\pi}{2}$$

$$= \frac{1}{2} [X(e^{j\frac{\pi}{4}}) + X(e^{j(\frac{\pi}{4} - \pi)})]$$

$$= \frac{1}{2} [X(e^{j\frac{\pi}{4}}) + X(e^{j(\frac{3\pi}{4})})]$$

$$= \frac{1}{2} \cdot X(e^{j\frac{\pi}{4}})$$